

IN THE CLAIMS:

Please amend claims 1- 13 as follows:

1. (Currently Amended) ~~An iterative~~ Iterative method for decoding a signal vector Y obtained from N sampled signals in a space-time communication system with M ~~transmit~~transmission antennae and N ~~receive~~receiving antennae, with N greater than or equal to M, with a view to obtaining an estimation of the symbols of the signals transmitted; characterized in that each iteration comprises the following steps:

- Pre-processing of the vector Y in order to maximize the signal to noise+interference ratio in order to obtain a signal \tilde{r}^ℓ ,
- Subtraction from the signal \tilde{r}^ℓ of a signal \hat{z}^ℓ by means of a subtractor, the signal \hat{z}^ℓ being obtained by reconstruction post-processing of the interference between symbols from the symbols estimated during the preceding iteration,
- Detection of the signal generated by the subtractor in order to obtain, for the iteration in progress, an estimation of the symbols of the signals transmitted; and in that, the N signals being processed by time intervals T corresponding to the time length of the linear space-time code associated with the transmitted signals, the pre-processing step involves the matrix B in order to maximize the signal to noise+interference ratio, the transfer function of which is:

$$B^\ell = \text{Diag} \left(\frac{1}{\rho_{\ell-1}^2 A_i^\ell + \frac{N_0}{E_s}} \right) \quad 1 \leq i \leq MT \quad C^H V^\ell$$

with $V^\ell = \begin{bmatrix} 1 - \rho_{\ell-1}^2 \cdot C \cdot C^H + Id_N \\ \frac{N_0}{E_s} \end{bmatrix}^{-1}$; $A^\ell = \text{diag}(C^H \cdot V^\ell \cdot C)$;

wherein ℓ : iteration index; ρ : standardized correlation coefficient between the real symbols and the estimated symbols; N_0 : noise variance; E_s : mean energy of a symbol; C : extended channel matrix;

and in that the post-processing step involves a matrix D for the reconstruction of the interference between symbols, the transfer function of which is:

$$D^\ell = B^\ell \cdot C \cdot \rho_{\ell-1} - \text{Diag} \left[\frac{1}{\rho_{\ell-1}^2 A_i^\ell + \frac{N_0}{E_s}} \right] \quad 1 \leq i \leq MT$$

2. (Currently Amended) Method according to claim 1, characterized in that ~~wherein~~ the pre-processing step is carried out by operating a matrix multiplication between the signal vector Y and a matrix B , the matrix B being updated at each iteration.

3. (Currently Amended) Method according to claim 1 or 2, characterized in that, ~~wherein~~ the post-processing step is carried out by operating a matrix

multiplication between the vector of the symbols estimated during the preceding iteration and the matrix D, the matrix D being updated at each iteration.

4. (Currently Amended) ~~Method~~ The method according to claim 2 or 3, characterized in that, wherein for each iteration, the standardized correlation coefficient ρ is calculated, the updating of a matrix being achieved by determining new coefficients of the matrix as a function of the correlation coefficient obtained for the preceding iteration.

5. (Currently Amended) ~~Method~~ The method according to any one of the preceding claims, characterized in that claim 1, wherein in order to determine the correlation coefficient ρ^ℓ for each iteration:

- the signal to interference ratio SINR is calculated using the following

$$\text{formula: } \text{SINR}^\ell = \left[\frac{1}{\xi^\ell e^{\xi^\ell} E_1(\xi^\ell)} - 1 \right] \frac{1}{1 - \rho_{\ell-1}^2}$$

and defining the integral exponential $E_1(s) = \int_s^{+\infty} \frac{e^{-t}}{t} dt$

$$\text{with } \xi^\ell = \frac{\varsigma}{1 - \rho_{\ell-1}^2} \text{ and } \varsigma = \frac{N_o}{N E_s}$$

- the symbol error probability Pr is calculated from the signal to interference ratio SINR^ℓ ; and

- the correlation coefficient ρ^ℓ is then calculated from the symbol error probability Pr.

6. (Currently Amended) ~~Method~~The method according to claim 5,
~~characterized in that, wherein~~ it is assumed that $\rho^0 = 0$.

7. (Currently Amended) ~~Method~~The method according to claim 5 or
6, ~~characterized in that, wherein~~ in order to calculate the symbol error probability Pr it is
assumed that the total noise is Gaussian.

8. (Currently Amended) ~~Method~~The method according to claim 7,
~~characterized in that, wherein~~ the formula corresponding to the constellation originating from
a linear modulation transmission is used.

9. (Currently Amended) ~~Method~~The method according to ~~any one of~~
~~claims 5 to 8, characterized in that, claim 5, wherein~~ in order to calculate the correlation
coefficient ρ^ℓ from the symbol error probability Pr , it is assumed that when there is an error,
the threshold detector detects one of the closest ~~neighbours~~neighbors to the symbol
transmitted.

10. (Currently Amended) ~~Method~~The method according to ~~any one of~~
~~the preceding claims, characterized in that, claim 1, wherein~~ at the final iteration, the signal
leaving the subtractor is introduced into a soft-input decoder.

11. (Currently Amended) ~~The method according to any one of~~
~~the preceding claims, characterized in that~~claim 1, wherein the information symbols are elements of a constellation originating from a quadrature amplitude modulation.

12. (Currently Amended) ~~Space-time~~A space-time decoder implementing a method according to ~~any one of the preceding claims~~claim 1 for decoding a signal vector Y obtained from N sampled signals in a space-time communication system with M ~~transmit~~transmission antennae and N ~~receive~~receiving antennae, with N greater than or equal to

M, with a view to obtaining an estimation of the symbols of the signals transmitted, characterized in that it comprises:

- a pre-processing module of the vector Y for maximizing the signal to noise+interference ratio in order to obtain a signal \tilde{r}^ℓ ,
- a subtractor for subtracting a signal \hat{z}^ℓ from the signal \tilde{r}^ℓ ,
- a post-processing module for the reconstruction of the interference between symbols from the symbols estimated during the preceding iteration in order to generate the signal \hat{z}^ℓ ,
- a threshold detector for detecting the signal generated by the subtractor in order to obtain, for the iteration in progress, an estimation of the symbols of the signals transmitted;

and in that the N signals being processed by intervals of time T corresponding

to the time length of the linear space-time code associated with the transmission signals, the pre-processing module consists of a matrix B for maximizing the signal to noise+interference ratio, the transfer function of which is:

$$B^\ell = \text{Diag} \left(\frac{1}{\rho_{\ell-1}^2 A_i^\ell + \frac{N_0}{E_s}} \right)_{1 \leq i \leq MT} \cdot C^H V^\ell$$

with $V^\ell = \begin{bmatrix} 1 - \rho_{\ell-1}^2 & C C^H + Id_N \\ \frac{N_0}{E_s} & \end{bmatrix}^{-1}$; $A^\ell = \text{diag} (C^H \cdot V^\ell \cdot C)$;

wherein ℓ : iteration index; ρ : standardized correlation coefficient between the real symbols and the estimated symbols; N_0 : noise variance; E_s : mean energy of a symbol; C : extended channel matrix;

and in that the post-processing module consists of a matrix D for the reconstruction of the interference between symbols, the transfer function of which is:

$$D^\ell = B^\ell \cdot C \cdot \rho_{\ell-1} - \text{Diag} \left(\frac{1}{\rho_{\ell-1}^2 A_i^\ell + \frac{N_0}{E_s}} \right)_{1 \leq i \leq MT}$$

13. (Currently Amended) ~~Decoder~~The decoder according to claim 12,
~~characterized in that~~wherein it comprises a soft input decoder receiving the signal originating
from the subtractor during the final iteration.